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FIELD OF INVENTION

The present invention relates generally to tunneling sensors and, more particularly, to a tunneling sensor with linear force rebalance and a method for fabricating the same.

BACKGROUND OF THE INVENTION

Some former force rebalance tunneling sensors used a single capacitor with a square law volts-to-force relationship. This yielded an output voltage proportional to the square root of the quantity to be measured. Alternative former force rebalance tunneling sensors used piezoelectric layers to perform the rebalance function.

Sensors providing a non-linear output are generally undesirable because they lead to harmonic distortion of the quantity being measured. Sensitivity also varies with the magnitude of input signals. Furthermore, the dynamic range over which a sensor yields a faithful representation of an input signal is reduced.

Piezoelectric rebalancing is generally inferior due to hysteresis, poor temperature and time stability, and small available displacements. The additional complexity of fabricating piezoelectric layers on a micromachined device is also undesirable.

On the other hand, linear force rebalancing increases dynamic range and reduces non-linearity, harmonic distortion, and

1 intermodulation distortion. For many applications, such as
2 phased arrays, linear operation is absolutely essential.

3 NASA's Jet Propulsion Laboratory (JPL) has designed a state
4 of the art tunneling accelerometer device primarily for use in
5 phased arrays (see "Tunnel-Effect Displacement Sensor", NASA Tech
6 Briefs, Vol. 13, No. 9, September 1989), but this device has
7 several minor drawbacks that may act as barriers to practical
8 use. For instance, the JPL device requires a high bias voltage.
9 Specifically, the JPL device currently requires a 200 volt bias
10 voltage to close the gap between the tunnel-effect tip. This
11 large voltage is necessary because of a large capacitor gap
12 (hundreds of microns) in the rebalance capacitor. This is an
13 uncommonly high voltage for use in towed arrays, inasmuch as high
14 voltages create corrosion and safety hazards in handling and
15 testing. Furthermore, the circuitry required to generate such a
16 high voltage can generate noise for the rest of the array.

17 Another drawback of the JPL device is that it employs non-
18 linear force rebalance. A single capacitor is used for force
19 rebalance in the JPL device, and the force across this single
20 capacitor is proportional to the square of the applied voltage.
21 This puts a non-linearity in the feedback loop wherein the output
22 voltage is proportional to the square root of the incident
23 acceleration. This, in turn, creates harmonic distortion,
24 intermodulation, and phase non-linearity, which leads to reduced
25 sensitivity and dynamic range. For array applications,

1 linearity, uniform phase, and low distortion are essential to the
2 combining of the numerous transducers which make up the array.

3 Still another drawback of the JPL device is its size, which
4 is on the order of 8 mm. This is fairly large for a
5 micromachined sensor. For many applications, such as thin line
6 towed arrays, this is simply too large.

7 Accordingly, it would be desirable to overcome the
8 disadvantages of former force rebalance tunneling sensors and
9 thereby provide a tunneling sensor having a pair of force
10 rebalance capacitors that are used in a push-pull relationship so
11 as to provide a rebalance force that is a linear function of
12 applied rebalance voltages, which leads to an output torque
13 voltage that is linearly related to input acceleration.

14 SUMMARY OF THE INVENTION

15
16 The present invention contemplates a tunneling sensor having
17 a pair of force rebalance capacitors that are used in a push-pull
18 relationship so as to provide a rebalance force that is a linear
19 function of applied rebalance voltages, which leads to an output
20 torque voltage that is linearly related to input acceleration.

21 The present invention tunneling sensor, which is constructed
22 primarily as a rotational accelerometer, comprises a plate
23 electrode that is formed from and attached to a silicon substrate
24 by a pair of torsional flexures, which provide an axis of
25 rotation for the plate electrode. A pendulous mass is formed on
26 a first end of the plate electrode, and a tunnel-effect contact

1 is formed on a second end of the plate electrode. A pair of
2 torque rebalance bridge electrodes are formed on the substrate so
3 as to span the plate electrode. A tunnel-effect tip is formed on
4 the substrate so as to be proximate the tunnel-effect contact and
5 in line with the rotational path that the tunnel-effect contact
6 takes when the plate electrode is rotated.

7 The plate electrode, and hence the tunnel-effect contact,
8 are typically grounded, while the pair of torque rebalance bridge
9 electrodes are complementarily driven with rebalance voltages,
10 having a constant bias voltage component and a output torque
11 voltage component, so as to generate an electrostatic rebalance
12 force that is a linear function of the rebalance voltages. A
13 small bias voltage is typically applied to the tunnel-effect tip
14 so as to induce the tunnel current. The result is an output
15 torque voltage that is linearly related to input acceleration.

16 Accordingly, the primary object of the present invention is
17 to provide a tunneling sensor having a pair of force rebalance
18 capacitors that are used in a push-pull relationship so as to
19 provide a rebalance force that is a linear function of applied
20 rebalance voltages, which leads to an output torque voltage that
21 is linearly related to input acceleration.

22 The above primary object, as well as other objects,
23 features, and advantages, of the present invention will become
24 readily apparent from the following detailed description which is
25 to be read in conjunction with the appended drawings.

1 BRIEF DESCRIPTION OF THE DRAWINGS

2 In order to facilitate a fuller understanding of the present
3 invention, reference is now made to the appended drawings. These
4 drawings should not be construed as limiting the present
5 invention, but are intended to be exemplary only.

6 Figure 1 is a plan view of a tunneling sensor with linear
7 force rebalance according to the present invention;

8 Figure 2 is a schematic block diagram of a control system
9 for the tunneling sensor shown in Figure 1;

10 Figure 3 is a cross-sectional side view, taken in relation
11 to line A-A of Figure 1, of a tunneling sensor according to the
12 present invention in its initial fabrication stage;

13 Figure 4 is a cross-sectional side view, taken in relation
14 to line A-A of Figure 1, of the tunneling sensor shown in Figure
15 3 after the initial oxide layer has been patterned on both its
16 front and back sides;

17 Figure 5 is a cross-sectional side view, taken in relation
18 to line A-A of Figure 1, of the tunneling sensor shown in Figure
19 4 after a boron diffusion has been performed on the silicon wafer
20 through the patterned openings in the initial oxide;

21 Figure 6 is a cross-sectional side view, taken in relation
22 to line A-A of Figure 1, of the tunneling sensor shown in Figure
23 5 after the initial oxide has been photolithographically removed
24 from certain regions thereof;

1 Figure 7 is a cross-sectional side view, taken in relation
2 to line A-A of Figure 1, of the tunneling sensor shown in Figure
3 6 after a sacrificial (spacer) layer has been deposited on
4 selected patterned regions;

5 Figure 8 is a cross-sectional side view, taken in relation
6 to line A-A of Figure 1, of the tunneling sensor shown in Figure
7 7 after a thin metal (seed) layer has been deposited thereon;

8 Figure 9 is a cross-sectional side view, taken in relation
9 to line A-A of Figure 1, of the tunneling sensor shown in Figure
10 8 after an electroplating mask has been deposited and patterned
11 thereon;

12 Figure 10 is a cross-sectional side view, taken in relation
13 to line A-A of Figure 1, of the tunneling sensor shown in Figure
14 9 after a pendulous weight, bridge electrodes, and a tunnel tip
15 have been electroplated in the open areas of the electroplating
16 mask;

17 Figure 11 is a cross-sectional side view, taken in relation
18 to line A-A of Figure 1, of the tunneling sensor shown in Figure
19 10 after the electroplating mask, the sacrificial (spacer) layer,
20 and the exposed portions of the thin metal (seed) layer have been
21 removed by polymer stripping;

22 Figure 12 is a cross-sectional side view, taken in relation
23 to line A-A of Figure 1, of the tunneling sensor shown in Figure
24 11 after an anisotropic EDP etch (ethylene-diamine, pyrocatechol,
25 and water) is preformed to substantially free up the plate
26 electrode.

1 DETAILED DESCRIPTION OF A PREFERRED EMBODIMENT

2 Referring to Figure 1, there is shown a plan view of a
3 tunneling sensor 10 with linear force rebalance according to the
4 present invention. The present invention tunneling sensor 10,
5 which is constructed primarily as a rotational accelerometer,
6 comprises a plate electrode 12 that is formed from and attached
7 to a substrate 14 by a pair of torsional flexures 16 that provide
8 an axis of rotation about which the plate electrode 12 is
9 rotatable. The plate electrode 12 and the pair of torsional
10 flexures 16 are typically etched out from the substrate 14, as
11 indicated by the etch slot 32 shown in Figure 1 (also see Figure
12 12) and as will be described in detail below.

13 A pendulous mass 18 is formed on a first end of the plate
14 electrode 12, and a tunnel-effect contact 20 is formed on a
15 second end of the plate electrode 12. A pair of torque rebalance
16 bridge electrodes 22 are formed on the substrate 14 so as to span
17 the plate electrode 12 (see Figure 12). A tunnel-effect tip 24
18 is formed on the substrate 14 so as to be proximate the tunnel-
19 effect contact 20 and in line with the rotational path that the
20 tunnel-effect contact 20 takes when the plate electrode 12 is
21 rotated. It should be noted that the pendulous mass 18 and the
22 tunnel-effect contact 20 may be formed on the same end of the
23 plate electrode 12 if such is adjusted for in the applied
24 rebalance voltages, as described in detail below.

25 The plate electrode 12, through one of the pair of torsional
26 flexures 16, each of the pair of torque rebalance bridge

1 electrodes 22, and the tunnel-effect tip 24 all have associated
2 electrically conductive contacts 26, 28, and 30, respectively, so
3 as to allow for wire bonding during packaging.

4 At this point it should be noted that the substrate 14 is
5 typically bulk silicon and the plate electrode 12 is typically a
6 boron diffused portion thereof, as is described in detail below.
7 The pair of torsional flexures 16 are included as part of the
8 boron diffused area, as indicated by the shaded area shown in
9 Figure 1. The pendulous mass 18, the pair of torque rebalance
10 bridge electrodes 22, and the tunnel-effect tip 24 are all
11 typically electroplated gold, or an electroplated gold layer
12 covered by another electroplated metal, as is also described in
13 detail below. The tunnel-effect contact 20 and all of the other
14 electrically conductive contacts 26, 28, and 30 are typically
15 formed of a bi-metal layer such as chrome-gold, as is further
16 described in detail below.

17 It should also be noted that the pair of torque rebalance
18 bridge electrodes 22 are preferably symmetrically spaced and
19 located over opposite ends of the plate electrode 12, equidistant
20 from the rotational axis provided by the pair of torsional
21 flexures 16, so as to facilitate in the rebalance force
22 linearization. Otherwise, a compensation voltage component must
23 be added to the applied rebalance voltages in order to achieve
24 linearization.

25 Referring to Figure 2, there is shown a schematic block
26 diagram of a control system 40 for the present invention

1 tunneling sensor 10 shown in Figure 1. The control system 40 has
2 a stiff feedback loop since the relative movement between the
3 tunnel-effect tip 24 and the tunnel-effect contact 20 is
4 preferably limited to $\pm 5 \text{ \AA}$ in use. Because of the high loop
5 gain required, stability is an issue, and compensation will be
6 required to ensure phase margin near the unity loop gain
7 frequency. In accordance with the present invention, the control
8 system 40 is shown having an input acceleration 42 and an output
9 torque voltage 44 that is linearly related to the input
10 acceleration 42.

11 The control system 40 comprises an acceleration to torque
12 conversion block 46 wherein the input acceleration 42 is
13 converted into torque as a result of the pendulous nature of the
14 plate electrode 12 and the pendulous mass 18. The acceleration
15 to torque conversion is essentially realized by multiplying the
16 weight of the pendulous mass 18 by the distance between the
17 center of gravity of the pendulous mass 18 and the axis of
18 rotation running through the pair of torsional flexures 16. The
19 converted torque is summed with torque that is produced as a
20 result of the electrostatic rebalance force between the pair of
21 torque rebalance bridge electrodes 22 and the plate electrode 12.

22 The control system 40 also comprises a plate rotation block
23 50 representing the differential equation modeling the mechanical
24 motion of the tunneling sensor 10 in response to an applied
25 torque. The coefficients of M , k_p , and k_θ are the total moment
26 of inertia of the plate electrode 12 and the pendulous mass 18, a

1 damping spring constant, and a rotational spring constant,
2 respectively. An angular to linear displacement conversion is
3 performed on the output of the plate rotation block 50. θ is the
4 angle of rotation of the plate electrode 12 in radians, and R_{tip}
5 represents the distance from the axis of rotation running through
6 the pair of torsional flexures 16 to the tunnel-effect tip 24.

7 A tip tunneling block 54 describes the current flow across
8 the gap between the tunnel-effect contact 20 and the tunnel-
9 effect tip 24. I represents the tunnel current, B represents a
10 bias voltage applied across the gap, α is a constant related to
11 the tunnel current, d is the linear displacement across the gap
12 (θR_{tip}), and Φ represents the potential barrier to the tunnel
13 current. The tip tunneling block 54 mathematically models the
14 current-voltage relationship at the tunneling tip.

15 The tunnel current, I , is converted into a representative
16 voltage, V , and a logarithmic amplifier 58 linearizes the
17 exponential dependence of the tunnel current, I , on the tip
18 displacement, d . A reference voltage, V_{ref} , corresponding to a
19 desired quiescent point for the control loop ($I \approx 1$ nA, $d \approx 5$ -10
20 Å) is summed with the output of the logarithmic amplifier 58 so
21 as to determine if any difference exists therebetween. The
22 resultant difference signal, if any exists, is passed through an
23 integrator 64 and a phase compensator 66 so as to provide the
24 output torque voltage 44 which is linearly related to the input
25 acceleration 42.

1 The force linearization block 68 utilizes the output torque
2 voltage 44 to generate the complementary rebalance voltages for
3 the pair of torque rebalance bridge electrodes 22. These
4 complementary rebalance voltages are produced by adding and
5 subtracting a constant bias voltage (V_{bias}) to the output torque
6 voltage (V_{torque}). These sum and difference voltages are then
7 applied to the pair of torque rebalance bridge electrodes 22 so
8 as to generate a rebalance torque against the plate electrode 12
9 that is proportional to $4V_{\text{bias}}V_{\text{torque}}$. Thus, the rebalance force is
10 linearly related to the output torque voltage 44, and hence to
11 the input acceleration 42. It should be noted that the voltage
12 level for the constant bias voltage (V_{bias}) is typically 10 VDC.

13 At this point it should be noted that the plate electrode
14 12, and hence the tunnel-effect contact 20, are typically
15 grounded, and a small bias voltage is typically applied to the
16 tunnel-effect tip 24. The voltage level for the bias voltage is
17 typically 0.2 VDC.

18 It should also be noted that the present invention tunneling
19 sensor 10 yields sensitivity on the order of 20 ng/ $\sqrt{\text{Hz}}$ at 1 kHz.
20 According to theoretical analyses, this is substantially more
21 sensitive than mere capacitive pickoffs at this frequency.

22 The method for fabricating the present invention tunneling
23 sensor 10 is in itself novel. Figures 3-12 show cross sections
24 of the tunneling sensor 10 at sequential stages of fabrication.

25 Referring to Figure 3, the tunneling sensor 10 is shown in
26 its initial fabrication stage comprising the silicon wafer

1 substrate 14 that is coated on both its front and back sides with
2 front 70 and back 72 dielectric layers, which may be silicon
3 dioxide, silicon nitride, or silicon carbide. The preferred
4 material for the dielectric layers 70 and 72 is thermally grown
5 silicon dioxide.

6 Referring to Figure 4, the tunneling sensor 10 is shown
7 after the front 70 and back 72 dielectric layers (hereinafter
8 referred to as the initial oxidation, or initial oxide, layers)
9 have been patterned using conventional photolithography and
10 either wet or dry etching.

11 Referring to Figure 5, the tunneling sensor 10 is shown
12 after a boron diffusion has been performed on selected regions
13 12, 76, and 78 of the silicon wafer 14 through the patterned
14 openings in the initial oxide layers 70 and 72. The initial
15 oxide layers 70 and 72 are used as a diffusion mask to
16 selectively diffuse boron through the patterned openings. The
17 boron diffusion is preferably carried out using a solid source
18 boron diffusion at a temperature between 1100°C and 1200°C,
19 although gas sources can also be used.

20 Referring to Figure 6, the tunneling sensor 10 is shown
21 after the initial oxide layers 70 and 72 have been
22 photolithographically removed from certain regions, such as etch
23 slot regions 80, and from the back side of the silicon wafer 14.
24 Also, a first bi-metal layer has been deposited by sputtering or
25 evaporating on selected patterned regions so as to form a tunnel-
26 effect contact 20 and various wire bond contacts, including those

1 for the plate electrode 26, the bridge electrodes 28, and the
2 tunnel-effect tip 30. This first bi-metal layer is preferably
3 chrome-gold, titanium-gold, or titanium/tungsten-gold.

4 Referring to Figure 7, the tunneling sensor 10 is shown
5 after a sacrificial (spacer) layer 74 has been deposited on
6 selected patterned regions. This spacer layer 74 may be
7 photoresist, polyimide, silicon dioxide, polysilicon, or other
8 sacrificial layers known to those skilled in the art. The
9 preferred spacer layer material is positive photoresist.

10 Referring to Figure 8, the tunneling sensor 10 is shown
11 after a thin metal (seed) layer 82 has been deposited (by
12 sputtering) over the entire front side of the wafer structure so
13 as to serve as a plating base for subsequent electroplating which
14 will form the bridge electrodes 22. This seed layer 82 must have
15 good adhesion to the various materials exposed on the front side
16 of the wafer structure, in addition to allowing easy
17 electroplating. Typically this seed layer 82 is formed of a bi-
18 metal deposit, with the first metal layer being chosen for good
19 adhesion to silicon dioxide and the second metal layer being
20 chosen for easy electroplating. The first metal (adhesion) layer
21 is typically titanium, chromium, titanium-tungsten alloy, or
22 aluminum. The second metal (electroplating) layer is typically
23 gold, chromium, copper, silver, nickel, palladium, or platinum.
24 A preferred embodiment uses a titanium-gold bi-layer as the
25 plating base (seed) layer 82.

1 Referring to Figure 9, the tunneling sensor 10 is shown
2 after an electroplating mask 84 has been deposited and patterned.
3 This mask 84 may be photoresist, e-beam resist, x-ray resist, or
4 polyimide. A preferred implementation uses a photoresist as the
5 plating mask 84.

6 Referring to Figure 10, the tunneling sensor 10 is shown
7 after the pendulous weight 18, the bridge electrodes 22, and the
8 tunnel-effect tip 24 have been electroplated in the open areas of
9 the electroplating mask 84. Gold is the preferred metal for the
10 electroplating, since gold is the preferred tunnel contact metal.
11 Alternatively, a thin gold layer may be electroplated first, and
12 a thicker layer of some other metal, such as nickel, silver, or
13 copper, may be electroplated thereon. It should be noted that
14 the bridge electrodes 22 have perforations formed therein so as
15 to reduce the damping spring coefficient, k_p .

16 Referring to Figure 11, the tunneling sensor 10 is shown
17 after the electroplating mask 84, the sacrificial (spacer) layer
18 74, and the exposed portions of the thin metal (seed) layer 82
19 have been removed by polymer stripping. The polymer stripping is
20 typically done in photoresist stripper, acetone, or by an oxygen
21 plasma. The portion of the seed layer 82 that is not protected
22 by the electroplated material 18 and 24 is stripped by an
23 appropriate wet or dry etch, such as are well known in the
24 industry.

25 Referring to Figure 12, the tunneling sensor 10 is shown
26 after an anisotropic EDP etch (ethylene-diamine, pyrocatechol,

1 and water) is preformed to substantially free up the plate
2 electrode 12. At this point, the tunneling sensor 10, which is
3 typically fabricated in an array of like sensors on the silicon
4 wafer 14, is ready for separation and packaging. Also shown is
5 the axis of rotation 86 of the plate electrode 12 running through
6 the pair of torsional flexures 16.

7 In view of the foregoing, it can be easily understood that
8 the present invention tunneling sensor 10 is smaller and easier
9 to use in common applications than the JPL device or similar
10 sensor devices. For example, the present invention tunneling
11 sensor 10 can easily fit on a 3 mm chip and can be used as an
12 accelerometer, a vibration sensor, a magnetic field sensor, a
13 pressure sensor, a hydrophone, and a microphone.

14 Also, the present invention tunneling sensor 10 requires
15 only moderate voltage levels (typically 20 volts) to achieve
16 rebalance and tip contact due to the small capacitor gaps
17 (typically 2 microns) used in surface micromachining.

18 The present invention is not to be limited in scope by the
19 specific embodiment described herein. Indeed, various
20 modifications of the present invention, in addition to those
21 described herein, will be apparent to those of skill in the art
22 from the foregoing description and accompanying drawings. Thus,
23 such modifications are intended to fall within the scope of the
24 appended claims. Additionally, various references are cited
25 throughout the specification, the disclosures of which are each
26 incorporated herein by reference in their entirety.